

GENERAL DESCRIPTION

CRG20 is the latest evolution of angular rate sensor utilising Silicon Sensing's patented cos20 ring structure MEMS technology. It is designed to meet the requirements of a wide range of automotive and commercial applications.

CRG20 builds upon the success of Silicon Sensing's benchmark family of MEMS rate sensors. Over 10 million of these have been delivered, with proven success in high performance automotive applications.

The sensor is a fully integrated digital solution, comprising MEMS sensing element, digital acquisition ASIC, and microprocessor. This single chip solution minimises the requirements for additional electronics, and thereby reduces space requirements and overall system costs. Fully digital closed loop control eliminates any temperature and ageing effects associated with analogue electronics, and provides exceptionally stable performance over a wide range of operating conditions.

The unique resonating ring structure makes the gyro inherently immune to high levels of shock and vibration.

In addition to a commanded self-test feature, CRG20 incorporates continuous self-testing of the complete operation of the sensor and the signal conditioning circuits. CRG20 has been designed to provide unparalleled sensor integrity, through the mitigation of potential error sources and false-plausible failure modes. System designers have the opportunity with CRG20 to eliminate the requirement for redundant sensors in high integrity applications.

The sensor provides a digital interface in the form of a SPI[®] port together with analogue output pins for customers who need to operate in the analogue domain. In addition, two auxiliary analogue input pins are available to digitise other sensors such as accelerometers or additional gyros; this enables multi-axis sensor clusters to be easily produced.

FEATURES

- Proprietary cos20 ring MEMS technology
- Out of plane (z-axis) rate response
- No requirement for calibration
- Continuous comprehensive self-test
- Suitable for high integrity applications
- High immunity to shock and vibration
- Very low angular random walk
- Wide dynamic range
- 0.03125 °/s digital resolution
- Digital SPI® interface
- Compact LCC SMT package
- Two auxiliary ADC inputs
- Ratiometric analogue rate outputs
- Internal temperature sensor
- RoHS compliant

APPLICATIONS

- Electronic Stability Control
- Vehicle rollover detection and control
- Navigation systems
- Inertial measurement units
- Adaptive cruise control
- Event data recorders / telematics
- Robotics
- Platform stabilisation

ORDERING INFORMATION

Model	Bandwidth	Digital Rate Range	Analogue Rate Range
CRG20-01	40 Hz	300 °/s	75 °/s
CRG20-02	75 Hz	300 °/s	300 °/s
CRG20-22	90 Hz	300 °/s	300 °/s

Factory programmable bandwidth and rate ranges mean that CRG20 is available in a choice of configurations. To discuss alternative options, please contact Silicon Sensing (contact information is detailed at the end of this document).

CRG20 Digital Angular Rate Sensor DSPM Induction sensori & trasduttori

CONTENTS

CONTENTS	2
ABSOLUTE MAXIMUM RATINGS	3
NORMAL OPERATING RATINGS	3
ESD SENSITIVITY	3
SPECIFICATIONS	4
PIN DIAGRAM	6
CRG20 PIN DESCRIPTIONS	6
RECOMMENDED CIRCUIT DESIGN	7
THEORY OF OPERATION	9
PRINCIPLE	9
IMPLEMENTATION	9
HIGH INTEGRITY RATE SENSING	9
STRUCTURE INTEGRITY	9
SYSTEM NODE OBSERVATION	
ANALOGUE FEEDBACK	
DSP VERIFICATION	
COMMANDED BIT	
SERIAL PERIPHERAL INTERFACE PORT	
EXTERNAL SYSTEM TO CRG20 MESSAGES	
EXTERNAL INTERFACE DISCRETE SIGNALS	
BIT_OK (OUTPUT)	
RESET_IN (INPUT)	
ANALOGUE RATE OUTPUT CHANNELS	
MECHANICAL DESIGN	
RECOMMENDED SOLDER CONDITIONS PACKAGE DETAILS	
PAURAGE DETAILS	15



ABSOLUTE MAXIMUM RATINGS

-40 °C to +130 °C -60 °C to +150 °C 6.0 V 75 mA <5 mV pk-pk in frequency range 100 Hz to 12 kHz <2 mV pk-pk in frequency range 12 kHz to 400 MHz

NOTE: Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or other conditions beyond those indicated is not implied.

6 V for 1 second with a current limit of < 100 mA

NORMAL OPERATING RATINGS

Reverse Power Supply Protection

Operating Temperature Supply Voltage Supply Current -40 °C to +105 °C 5 V dc nominal, (4.75 V min, 5.25 V max continuous) 60 mA maximum @ 5.25 V across operating temperature

ESD SENSITIVITY

The CRG20 device is rated to 2 kV using the Human Body Model (direct contact, 100 pF/1.5 kΩ).



ESD (Electrostatic Discharge) Sensitive Device This product may be damaged by ESD. Such damage may result in subtle performance or life degradation, or complete device failure. Appropriate storage and handling precautions should be taken at all times to avoid ESD damage.

FUNCTIONAL BLOCK DIAGRAM

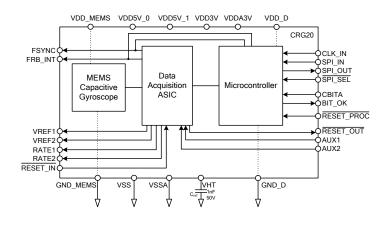


Figure 1



SPECIFICATIONS

Ambient Temperature (T_A) = -40 °C to 105 °C, V_{DD5V} = 4.75 V to 5.25 V, Angular Rate = 0 %, unless otherwise stated.

Table 1

Digital interface - Scale Pactor 20-01 0.002 20-02 0.002 0.003125 Scale factor and inearity Over full operating temperature range % 4.2.00 5.2.0.0.0.0.0.0.0.0.0.0.0.0.0.0.0.0.0.0.	Parameters	Conditions	Units		Value *1	
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Rate1 & Rate2 reference voltage Rate1 = Rate2 = V _{REF1} at 0 °/s V _{REF1} Cross Coupling Errors	Rate1 & Rate2 scale factor error	With respect to V _{REF1}	%		±0.5	
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Axis misalignment stability mrad ±8 Cross axis sensitivity % ±2 Acceleration induced misalignment mrad/(m/s²) 0.2 Dynamic Response			mrad	1	±17	
Cross axis sensitivity % ±2 Acceleration induced misalignment mrad/(m/s ²) 0.2 Dynamic Response						
Dynamic Response			%		±2	
	Acceleration induced misalignment		mrad/(m/s ²)		0.2	
	Dynamic Response			-		
	Gain peaking		dB		<1	

CRG20 Digital Angular Rate Sensor

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Page 4

CRG20 Digital Angular Rate Sensor

Parameters	Conditions	Units		Value *1	
			CRG 20-01	CRG 20-02	CRG * ⁷ 20-22
Bandwidth	-3 dB point	Hz	≥40	≥75	≥90
Angular acceleration limit		°/s²		>10,000	
Noise					
Over 1 to 25 Hz band		°/s rms, typ.	0.20	0.30	0.30
In-band noise	Up to the bandwidth of the variant	°/s rms, typ.	0.33	0.80	1.00
Over 1 to 101 Hz band		°/s rms, typ.	0.57	1.00	1.00
Broadband noise		°/s rms, typ.	0.70	1.40	1.50
Angular random walk		°/√hr		<1	
Commanded BIT (CBITA)					
Nominal rate output offset	T _A = 25 °C	°/s		+10.0	
Offset variation/tolerance	Between -40 °C and 85 °C	°/s		±3.00	
Offset variation/tolerance	Between 85 °C and 105 °C	°/s		±3.75	
Max data loss during test		ms		200	
Power Supply					
V _{DD_D}					
V _{DD5V_0}		V		·5 ±0.25 *	-5
V _{DD5V_1}		v	-	J 10.25	
V _{DD_MEMS}					
V _{DD3V} * ⁶		V		+3.3 ±0.2	
V _{DDA3V}		V	+3.3	(15 mA r	max)
V _{HT}	Generated internally by CRG20	V		+28 ±2	
V _{DD} Quiescent supply current	V _{DD5V} = 5.25 V	mA		60	
Voltage References					
V _{REF1}		V	V _{DD}	_{5V} /2 ± 0.2	5 %
V _{REF1} Voltage offset error		mV		±14 max	
V _{REF1} Voltage stability over temperature and life		mV		±7.5	
V _{REF1} Load regulation	Sinking < 0.03 mA into V_{REF1} pin, sourcing <1 mA from V_{REF1} pin	%		±0.5	
V _{REF1} Output noise	< 200 Hz < 20 kHz	mV rms		0.2 4	
V _{REF1} Load capacitor required		nF		1 ±0.2	
V _{REF2}		V	V	$\frac{120.2}{005}$ ±1	%
V_{REF2} Voltage stability over temperature and life		mV		±7.5	/0
V_{REF2} Load capacitor required	Low ESR Tantalum type $(0.1 - 0.5 \Omega)$	μF		10 ±2	
Digital Signals					
Voltage level		14	E		aio
		V	5 V	CMOS lo	gic
Input leakage current, I _{IN} Input capacitance, C _{IN}		nA pF		50 max 10 max	
Start-Up					
Start-up time		ms		<500	
			1		
Environment Temperature		°C	ļ	40 to +10	5
Shock	¹ / ₂ sine, t = 2 ms, recovery within 20 ms			95 min	
Vibration	20 Hz to 2 kHz	g g rms		8.85	
Mass				0.8	
191055		g		0.0	
Mechanical Characteristics		2222 /L		74	
Package coefficient of thermal expansion		ppm/K		7.1	

*1 Unless otherwise stated errors quoted in this table refer to the digital rate output.

*² Measurements taken following initial 2 hour powered soak at 105°C then after 30 minute dwell at the following temperature points (°C): 125, 105, 85, 50, 25, 0, -20, -40

*³ Where pin AUX1 is mentioned, the parameters are also applicable to AUX2.

*⁴ The maximum measurable input bandwidth is limited by the communication scheme in the SERIAL PERIPHERAL INTERFACE PORT section, i.e. 500 Hz (half the maximum message rate).

*⁵ Operating voltage stated as 4.75 V-5.25 V (5 V optimum). Device to be connected as per Figure 3.

*6 V_{DD3V} is generated internally, up to 15 mA maximum draw load may be applied externally.

*7 All CRG20-22 results are provisional pending qualification



PIN DIAGRAM

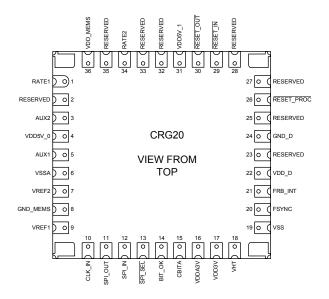


Figure 2

CRG20 PIN DESCRIPTIONS

Pin Name	Pin No	Pin Type¹	Expected Value	Description
				Power Supply Connections
VDD5V_0	4		+5 V ±0.25 V	Analogue 5 V power supply connection 0. Used to power elements of the ASIC in the CRG20.
VDD5V_1	31	I	+5 V ±0.25 V	Analogue 5 V power supply connection 1. Used to power elements of the ASIC in the CRG20, including the charge pump for the high tension (HT) voltage. It is recommended that this connection be routed directly to the pin without passing through other power supply pins to minimise the effect of the charge pump operation on the CRG20.
VDD3V	17	Ι	+3.3 V ±0.2 V	3.3 V power supply connection for the ASIC. VDD3V does not need to be supplied with 3.3 V as the CRG20 contains an internal 3.3 V regulator.
VDDA3V	16	0	+3.3 V	Connection to the CRG20 internal 3.3 V regulator. This can be used to supply 3.3 V at currents up to 15 mA.
VDD_D	22	-	+5 V ±0.25 V	Digital 5 V power supply connection for the microcontroller in the CRG20.
VDD_MEMS	36	-	+5 V ±0.25 V	Power supply connection for the MEMS and pickoff amplifier in the CRG20.
VHT	18	0	+28 V	Decoupling connection for the HT voltage charge pump in the CRG20. A 22 nF capacitor should be connected between VHT and VSS. Due to the sensitive nature of the HT voltage, the 22 nF capacitor should be on the same side of the mounting board as the CRG20 and the tracks should be as short as possible and not pass through any vias.
VSS	19	G	0 V	Ground connection for parts of the ASIC in the CRG20. It is recommended that the return path for this pin is not routed through any other ground pins, to minimise the effect of the charge pump operation on the CRG20.
VSSA	6	G	0 V	Ground connection for parts of the ASIC in the CRG20.
GND_D	24	G	0 V	Ground connection for the microcontroller in the CRG20.
GND_MEMS	8	G	0 V	Ground connection for the MEMS in the CRG20.
				Analogue Signals
AUX1	5	I	(VSSA+0.25 V) to (VDD5V-0.25 V)	External analogue input channel 1. The signal is digitised using a 12-bit analogue to digital converter (ADC) and can be observed using the SPI communications. See "Serial Peripheral Interface Port" section for details.
AUX2	3	I	(VSSA+0.25 V) to (VDD5V-0.25 V)	External analogue input channel 2. The signal is digitised using a 12-bit ADC and can be observed using the SPI communications. See "Serial Peripheral Interface Port" section for details.
RATE1	1	0	VREF1 ±2.3 V max	Buffered analogue output 1. This signal represents the angular rate applied to the CRG20 and is referenced to VREF1.
RATE2	34	0	VREF1 ±2.07 V max	Buffered analogue output 2. This signal represents the angular rate applied to the CRG20 and is referenced to VREF1. RATE2 is set to VSS if a continuous or demanded built in test (BIT) failure occurs.
VREF1	9	0	VDD5V / 2 ±7.5 mV	Analogue reference output 1. This output is used as the 0 °/s reference for RATE1 and RATE2.
VREF2	7	0	VDD5V / 3 ±7.5 mV	Analogue reference output 2. VREF2 is used for internal referencing only and is brought out to a pin for the required decoupling. See "Recommended Layout" section for details.
				Digital Signals
BIT_OK	14	0	5 V CMOS Logic	The CRG20 performs continuous self-checks and outputs the gross operational status of the device via the BIT_OK flag. If the applied angular rate exceeds the capability of the CRG20 and causes the output to fold back then a BIT error will be flagged.
CBITA	15	I	5 V CMOS Logic	The CRG20 has the ability to initiate a commanded built in test. This can be done using the SPI

CRG20 Digital Angular Rate Sensor

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Page 6

Via Paolo Uccello 4 - 20148 Milano

CRG20-00-0100-110 Rev 13 DCR710016657

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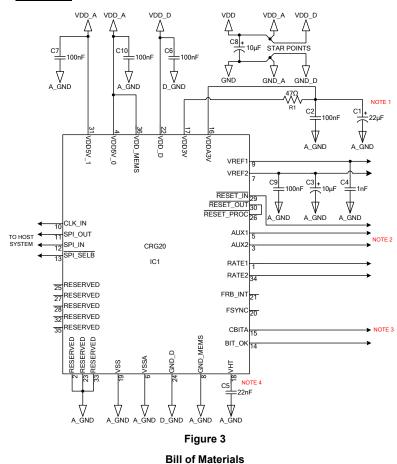


Pin Name	Pin No	Pin Type ¹	Expected Value	Description			
				communications (see "Serial Peripheral Interface Port" section for details) or the CBITA pin.			
				CBITA is primarily used as a 'quick test' to check the functionality of the device. While CBITA is			
				asserted, an offset is added to the rate sensing loop. BIT_OK shows a failure while CBITA is			
				active.			
CLK_IN	10	Ι	5 V CMOS Logic	SPI clock signal. This signal is typically a 1 MHz clock. See "Serial Peripheral Interface Port" section for details.			
				Frequency pulse output synchronous to the MEMS gyroscope resonant frequency. This signal			
FRB_INT	21	0	5 V CMOS Logic	will be in the range 3.35 kHz to 3.65 kHz. If the signal is used it should be buffered to reduce the			
				risk of affecting the operation of the CRG20.			
				Frequency output synchronous to 1024 times the MEMS gyroscope resonant frequency. This			
FSYNC	20	0	5 V CMOS Logic	signal will be in the range 13.7216 MHz to 14.9504 MHz. If the signal is used it should be			
				buffered to reduce the risk of affecting the operation of the CRG20.			
RESET IN	29	1	Active Low	When activated with an active low pulse, the ASIC is reset and 20 µs later an active low reset			
		/ totive Low	signal is sent to RESET_OUT.				
				RESET_OUT effectively mirrors the signal properties of RESET_IN with a 20 μ s delay.			
RESET_OUT	30	0	Active Low	RESET_OUT is designed to be connected to RESET_PROC to reset the processor following an			
				ASIC reset.			
				When activated with an active low pulse, the processor is reset to initial conditions. The ASIC is			
RESET PROC	26	1	Active Low	not reset. The recommendation is to connect RESET_PROC to RESET_OUT so that the			
	20			processor and ASIC are reset at the same time (effectively resetting the device). See			
				"Recommended Layout section for details).			
SPI_IN	12		5 V CMOS Logic	SPI data input (MOSI). See "Serial Peripheral Interface Port" section for details.			
SPI_OUT	11	0	5 V CMOS Logic	SPI data output (MISO). See "Serial Peripheral Interface Port" section for details.			
SPI_SEL	13		Active Low	SPI chip select. This line is active low. See "Serial Peripheral Interface Port" section for details.			
				Miscellaneous Signals			
RESERVED	2	-	0 V	Reserved for use by Silicon Sensing Systems Ltd. Should be tied to 0 V.			
RESERVED	23	-	0 V	Reserved for use by Silicon Sensing Systems Ltd. Should be tied to 0 V.			
RESERVED	25	-	-	Reserved for use by Silicon Sensing Systems Ltd. Should be left unconnected.			
RESERVED	27	-	-	Reserved for use by Silicon Sensing Systems Ltd. Should be left unconnected.			
RESERVED	28	-	-	Reserved for use by Silicon Sensing Systems Ltd. Should be left unconnected.			
RESERVED	32	-	-	Reserved for use by Silicon Sensing Systems Ltd. Should be left unconnected.			
RESERVED	33	-	0 V	Reserved for use by Silicon Sensing Systems Ltd. Should be tied to 0 V.			
RESERVED	35	-	-	Reserved for use by Silicon Sensing Systems Ltd. Should be left unconnected.			

¹ Pin types include: Input (I), Output (O) and Ground (G).



Schematic



No.	Ref. Des.	Value	Voltage	Tolerance	Package
1	IC1	CRG20	-	-	LCC-36
2	C1	22 µF	> 7 V	≤ 10 %	Case C
3	C2	100 nF	≥ 16 V	≤ 10 %	0603
4	C3	10 µF	≥ 16 V	≤ 10 %	Case B
5	C4	1 nF	≥ 16 V	≤ 10 %	0603
6	C5	22 nF	≥ 50 V	≤ 10 %	0603
7	C6	100 nF	≥ 16 V	≤ 10 %	0603
8	C7	100 nF	≥ 16 V	≤ 10 %	0603
9	C8	10 µF	≥ 16 V	≤ 10 %	Case B
10	C9	100 nF	≥ 16 V	≤ 10 %	0603
11	C10	100 nF	≥ 16 V	≤ 10 %	0603
12	R1	47 Ω	≥ 16 V	≤1%	0603

Table 2

Schematic Notes

- Note 1: Capacitor C1 should be chosen to have an effective series resistance (ESR) of between 0.1 Ω and 0.5 $\Omega.$
- Note 2: If the auxiliary inputs are not used they should be tied to a voltage between GND and VDD such as VREF1 (effectively midrail). This will prevent the AUX inputs causing ADC overflow errors which would result in a BIT failure.
- Note 3: Pin 15 (CBITA) of IC1 must be pulled down to GND to avoid the command BIT sequence triggering erroneously.
- Note 4: Pin 18 (VHT) of IC1 is a high impedance, high DC voltage signal and is sensitive to leakage currents. For this reason, the track connecting to C5 from pin 18 of IC1 should not contain any vias (i.e. the capacitor should be on the same side as the CRG20).

General PCB Design Notes

- IC1 is a ceramic package with a coefficient of thermal expansion (CTE) of 7.1 ppm/K. Board materials shall be chosen accordingly to minimise thermal mismatch.
- A ground screen is not shown to improve clarity. While a screening layer is beneficial, care must be taken to minimise copper as this increases the CTE of the board (see above).



Page 7 Via Paolo Uccello 4 - 20148 Milano

CRG20-00-0100-110 Rev 13 DCR710016657

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CRG20 Digital Angular Rate Sensor

Hatched screens may be used to minimise this effect.

- ¹/₂ ounce copper is recommended.
- If components are to be mounted on both sides of the board, care shall be taken to minimise the track lengths connecting decoupling capacitors to IC1, especially the capacitor C7 between pins 19 and 31.
- The tracks labelled 'VDD' and 'GND' shall be connected to the system 5 V and ground, respectively. Care shall be taken to minimise the resistance and inductance of these tracks, hence these tracks shall be as short as possible.
- The recommended package size for components has been chosen for ease of availability and is not restricted providing the values and tolerances are met.

PCB Layout

The recommended PCB layout shown in Figures 4 to 9 has been tracked on 2 layers for simplicity. Higher numbers of board layers can be used for screening, although attention must be paid to the CTE of the package and matching it to that of the PCB material. Copper layers increase the CTE of FR4. Figure 4 shows the different Gerber data layers stacked together to represent the entire PCB.

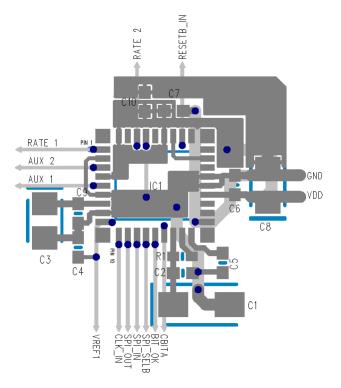
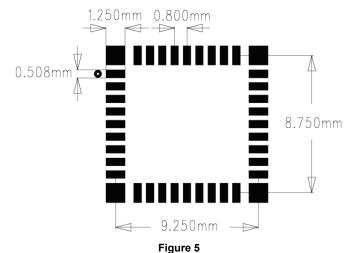
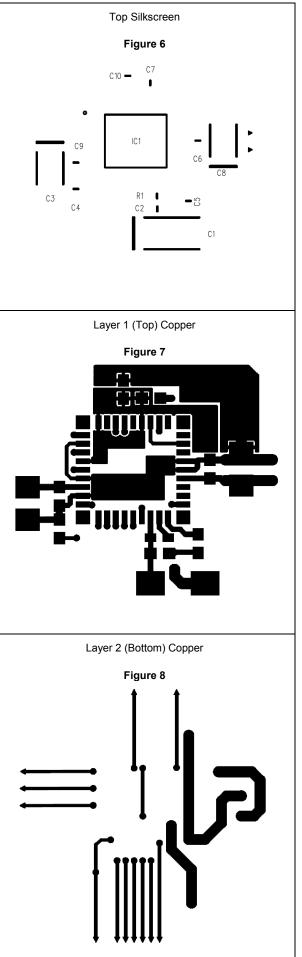


Figure 4

The recommended pad layout for the CRG20 36 pad LCC package is shown in Figure 5 and has been designed according to IPC 7351.





CRG20 Digital Angular Rate Sensor

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Page 8

CRG20-00-0100-110 Rev 13 DCR710016657



THEORY OF OPERATION

The CRG20 uses a bulk silicon Micro-Machined ring structure, a mixed signal ASIC and a processor to make a single axis yaw rate sensor.

Principle

The Silicon ring is driven in a $\cos 2\theta$ mode shape to produce the radial velocity components required to make a Coriolis gyroscope.

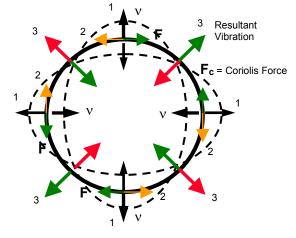


Figure 9

Figure 9 shows the movement of the silicon ring while vibrating. As the ring oscillates in its natural state the ring moves in the way shown by (1). When the gyro is turned, the Coriolis force acts on the ring as shown by (2). This causes a resultant vibration 45° out of alignment with the primary vibration, as shown by (3).The force required to completely null this resultant vibration is directly proportional to the angular rate.

Implementation

Eight uniformly spaced capacitive transducers are placed around the ring to form 2 pairs of drive transducers and 2 pairs of pickoff transducers.

One pair of diametrically opposed drive transducers is used to excite the $cos2\theta$ mode with the phase and amplitude sensed by the corresponding pickoff transducer pair. The drive amplitude and frequency is controlled by the ASIC and Processor to establish an accurate radial velocity component at the resonant frequency of the structure. This sets the basic operating point and scale factor of the system.

A secondary rate nulling loop is set up using the other drive and pickoff transducer pairs to enable the Coriolis forces (generated by applied angular rate) to be detected and servoed to zero using digital filters within the processor. This fully closed loop operation largely removes dependency of performance on the mechanical Q of the resonator and enables a very accurate level of bias and scale factor performance to be achieved without any compensation. The force required to null the coriolis force becomes an accurate measure of the applied angular rate.

All loop controllers and filtering are implemented within the processor so the rate (and other information) is available directly in digital form. This eliminates performance variation or drift in more typical analogue control loops, as the numeric control is inherently free from ageing, temperature and manufacture tolerances. All numeric processing is synchronised to the MEMS resonant frequency, so as to eliminate phase related detector errors commonly found in MEMS sensors.

The derived rate data is digitally filtered, scaled and then output via the SPI bus.

For legacy applications, the rate data is also converted to analogue format (with factory programmable scale factor and bias) and output on RATE1 and RATE2.

The digital rate data is output on the SPI interface together with other data (see SERIAL PERIPHERAL INTERFACE PORT) to produce a high integrity system.

HIGH INTEGRITY RATE SENSING

A passive BIT system continuously monitors the health of the CRG20, incorporating the following techniques.

Structure Integrity

The structure has been designed to 'fail silent' such that, in the unlikely event of the MEMS suffering mechanical damage, it will immediately stop vibrating as the changes in the mechanical tuning of the structure will cause significant change to the ring resonant frequency. Also, as there is only one interface to the structure, it isn't possible to lose part of it and continue to operate with reduced sensitivity.

System Node Observation

Due to the digital implementation of the control loops, every node can be monitored within the DSP code without additional complex circuitry or time delays. The subsequent reporting forms part of the standard payload of the SPI interface, available within every message, at one millisecond intervals

Analogue Feedback

The digital rate output is converted to an analogue voltage by passing through the internal SPI bus from the DSP to the ASIC. This value is converted in the same DAC as is used for force-rebalance. Two separate checks are now possible. Firstly, the analogue rate can be digitised by a host system, and compared with the rate as reported over the SPI bus. Secondly, this analogue signal can be connected to an auxiliary input of the same ADC as is used to monitor the transducers. It is redigitised within the sensor and presented to the host as an auxiliary data field in the SPI payload. Consequently, there are three measures of yaw rate that can be compared – providing triple verification.

DSP Verification

Conventional systems using a DSP have a residual failure mode whereby corruption of the arithmetic unit could cause the sensor to report an incorrect rate output with no means of internally validating the result. In CRG20, this is prevented by utilising time variant data to check for a predictable result to a test calculation. Using the low order byte of rate data from the previous message, a test calculation is run that exercises all of the operation types required to implement the control loop filters, and a packet of data is generated within the next SPI message payload (rate check byte). This is verified by the host, simply by implementing the same calculation, confirming the ALU integrity of both the CRG20 and the host.

COMMANDED BIT

The failure identification mechanisms described in the preceding sections operate continuously without interrupting or modifying the rate output. This is regarded as Continuous or Background Built In Test (BIT). There is also a commanded BIT, which is provided for backwards compatibility with legacy systems. In response to an external command, (either via the SPI interface or in response to a standard logic high level applied to the "CBITA" input), the secondary loop is disturbed to artificially invoke a fixed rate offset. This causes the output to change by a predetermined offset, which can be detected by the host system either via the SPI interface or via the analogue output.

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SERIAL PERIPHERAL INTERFACE PORT

Communication with the CRG20 may use a serial peripheral interface (SPI) port, in which the CRG20 behaves as the slave. When this port is used, at certain temperatures there is a risk of an interaction between the SPI timing and the resonator frequency. This interaction is seen as a beat frequency on the digital rate output. If the analogue rate output is used alone, this interaction will not occur.

The SPI port includes four signals: SPI select (SPI_SEL), serial clock in (CLK_IN), MOSI or SPI data in (SPI_IN) and MISO or SPI data out (SPI_OUT). The SPI_SEL line enables the CRG20 SPI port and frames each communication. When the SPI_SEL line is high, SPI_OUT is held in a high impedance state and signals on CLK_IN and SPI_IN are ignored by the CRG20. A communication consists of 48-bit data frames split into 6 8-bit bytes. The SPI port operates in full duplex mode, which means that as data is transmitted to the CRG20 on SPI_IN, it is also received from the CRG20 on SPI_OUT.

Table 3 shows detailed timing information referenced from Figure 10 and Figure 11, which shows the detail of the SPI port data structure.

Table 3 - SPI Port Timing Information

Parameter	Min	Max	Unit	Description
	Value	Value		
t _{CLK}	0.8		μs	CLK_IN period.
t ₁	5		μs	SPI_SEL to CLK_IN setup time.
t ₂	0.4 х t _{ськ}		μs	CLK_IN high pulse width.
t ₃	0.4 х t _{ськ}		μs	CLK_IN low pulse width.
t ₄	1.1		μs	Delay between successive bytes.
t _Q	730		μs	Minimum quiet time required between SPI_SEL rising and start of next communication (Note 1)
t _h	0		μs	SPI_SELhold time after final clock
t _m	53.5	270	μs	message duration (note 1)
t _{tot}	950		μs	repetition rate (Note 1)

Note 1: A quiet time (tq) between messages is required, and t1 + Tm + Th + tq = Ttot. Optimum performance is achieved with a repetition rate of 1 ms. Select an appropriate Tclk to meet the requirements of Tm, and then adjust tq to give a Ttot of 1 ms.

EXTERNAL SYSTEM TO CRG20 MESSAGES

The data frames from the external system to the CRG20 are in the following format:

Byte	Function
0	Control Byte
1	Reserved
2	Reserved
3	Reserved
4	Reserved
5	Checksum

Control Byte:

RES	RE	S	RES	RES	CBITA	NMT2	NMT1	NMT0		
bit 7								bit 0		
bit 7-4	•	RES Set	3 : Reser to 0	ved						
bit 3	:	CBI	TA : Con	nmandeo	d Built In	Test A				
			Deactiva	_	-					
		1 = /	Activate	CBIT_A	test (see	CBITA	function)			
bit 2-0	:	NM	Γ2:NMT0): Next N	Nessage	Туре				
		000	= Basic	rate data	a (default)				
		001	= Rate /	Aux1 sei	nsor data	1				
		010 = Rate / Aux2 sensor data								
		011	= Rate /	Temper	ature ser	nsor data	a			
		100	= Device	e configu	ration 1	data				
		404	Decide			-1-1-				

- 101 = Device configuration 2 data
- 110 = Reserved
- 111 = Reserved

Reserved:

RES	R	ES	RES	RES	RES	RES	RES	RES
bit 7	1 1	20	INLO	INEO	INEO	INLO	INLO	bit 0
								DILU
h:+ 7 0		DE	•	ام م ا				
bit 7-0		RE:	S : Rese	rvea				

Set to '00' hex

Checksum:

CHK7	CHK6	CHK5	CHK4	CHK3	CHK2	CHK1	CHK0
bit 7							bit 0

bit 7-0 : CHK7:CHK0 : Checksum

1s complement of the sum of bytes 0 - 4

The checksum is the 1s complement of sum of the previous 5 bytes

CRG20 to External System Messages

The data frames transmitted from the CRG20 to the external system are in the following format:

Byte	Function
0	Status Byte
1	Data 0
2	Data 1
3	Data 2
4	Data 3
5	Checksum

Each message from the CRG20 begins with a status byte contained in byte 0 and a checksum in byte 5.

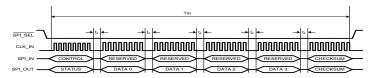


Figure 10 - SPI Data Frame Structure

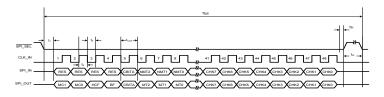


Figure 11 - SPI Data Timing Diagram

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Status	Bvte:	

Status I	Byte:						
MC1	MC0	HCF	BF	CBITA	MT2	MT1	MT0
bit 7							bit 0
bit 7-6	Inci	1:MC0 : remented	l by the (CRG20 e	ach time	e the hos	t
bit 5	HC 0 =	tem take F : Host No fault	commun detected	ications f			
bit 4	BF 0 =	Communication: Built in Gyro our BIT Failu	test (BIT tput OK		ected		
bit 3	: CB 0 =	ITA : CB CBIT_A	ITA activ test inac	tive			
bit 2-0 Checks	: MT : 000 001 010 011 100 101	1 = CBIT_A test in progress MT2:MT0 : Message type 000 = Basic rate data (default) 001 = Rate /Aux1 sensor data 010 = Rate / Aux2 sensor data 011 = Rate / Temperature sensor data 100 = Device configuration 1 data 101 = Device configuration 2 data					
Checks	um:						
CHK7	CHK6	CHK5	CHK4	CHK3	CHK2	CHK1	CHK0

CHK7	CHK6	CHK5	CHK4	CHK3	CHK2	CHK1	CHK0
bit 7							bit 0

CHK7:CHK0 : Checksum bit 7-0 : 1s complement of the sum of bytes 0 - 4

Depending on the message type requested in NMT2:NMT0 in the previous control byte from the host, the data bytes (Data 3: Data 0) in the subsequent data frame from the CRG20 contain the following data, indicated by NMT2:NMT0:

Basic Rate Data Message (NMT2:NMT0 = 000)

Byte	Function	Description
0	Status Byte	
1	Rate MS	Nominally ±350 °/s,
2	Rate LS	0.03125 °/s/lsb
3	Rate Check Byte	See Page 9
4	Reserved	Reserved
5	Checksum	

Rate / Aux1 Message (NMT2:NMT0 = 001)

Byte	Function	Description
0	Status Byte	
1	Rate MS	Nominally ±350°/s,
2	Rate LS	0.03125°/s/lsb
3	Aux1 MS	Aux1 data (signed integer),
4	Aux1 LS	1.22 mV/lsb
5	Checksum	

Rate / Aux2 Message (NMT2:NMT0 = 010)

Byte	Function	Description
0	Status Byte	
1	Rate MS	Nominally ±350 °/s,
2	Rate LS	0.03125 °/s/lsb
3	Aux2 MS	Aux2 data (signed integer),
4	Aux2 LS	1.22 mV/lsb
5	Checksum	

Rate / Temperature Sensor Message (NMT2:NMT0 = 011)

Byte	Function	Description
0	Status Byte	
1	Rate MS	Nominally ±350 °/s,
2	Rate LS	0.03125 °/s/lsb
3	Temperature MS	-50 °C to 145 °C, 0.125 °C/lsb
4	Temperature LS	-50 C to 145 C, 0.125 C/ISD
5	Checksum	

Device Configuration 1 Message (NMT2:NMT0 = 100) -Default on power up

Byte	Function	Description
0	Status Byte	
1	Bits (7:4) Bits (3:0)	Model rate range (e.g. 3 = 75 °/s) Model bandwidth (e.g. 1 = 100 Hz)
2	Bits (7:4) Bits (3:0)	Spare bits Model variant
3	Software Version	Software variant number
4	Bits (7:5) Bits (4:0)	Spare bits Year of manufacture
5	Checksum	

Device Configuration 2 Message (NMT2:NMT0 = 101)

Byte	Function	Description
0	Status Byte	
1	Bits (7:4)	Month of manufacture (3:0)
I	Bits (3:0)	Package lot number (11:8)
2	Bits (7:0)	Package lot number (7:0)
3	Bits (7:6)	Assembly plant (1:0)
3	Bits (5:0)	Device serial number (13:8)
4	Bits (7:0)	Device serial number (7:0)
5	Checksum	

Note: The contents of the two device configuration messages give each sensor a unique serial number.

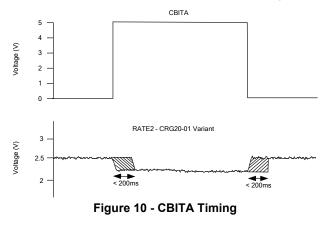
Note: If an incorrect message is received, the CRG20 will continue to transmit the message chosen in the last valid request. Upon power up, the CRG20 will output the device configuration 1 message until a valid control signal is received.

EXTERNAL INTERFACE DISCRETE SIGNALS

Two discrete signals are part of the external host interface of the CRG20:

CBITA (Input)

The CRG20 has the ability to initiate a commanded built in test. This is used as a test to ensure the device is functioning correctly, by exercising the CRG20 MEMS structure, the rate sensing loops and drive and pickoff electronics. CBITA works by applying an offset to the loops running inside the rate sensor, which results in a 10 °/s offset bias in the rate output. The timing of the offset in relation to the state of the CBITA pin is shown in Figure 10.



CBITA Logic Levels

-	
CBITA = 0	The CBIT/
CBITA = 1	The CRG
	which a n
	he annlier

A test will be disabled. 20 will initiate a CBITA test, in nominal offset bias of 10 °/s will be applied to the rate output.



BIT_OK (output)

While operational, the CRG20 performs continuous selfchecks during operation and outputs the gross operational state of the CRG20 with the BIT_OK flag.

The built in test (BIT) function is a comprehensive test of the sensor and the signal conditioning circuits. BIT_OK can take up to 500 ms to show a valid rate output after the CRG20 is powered up, as shown in Figure 11.

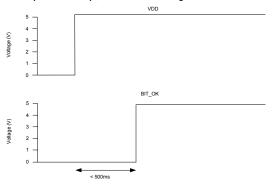


Figure 11 - BIT_OK Timing

During normal operation, the BIT_OK flag is set within 25 ms of a fault condition.

BIT_OK Logic Levels

BIT_OK = 0 The rate output from the CRG20 is invalid, as the built in test has failed (or CBIT is running)

BIT_OK = 1 The rate output is valid.

If the applied angular rate exceeds the capability of the CRG20 and causes the rate output to fold back then a BIT error will be flagged.

RESET_IN (input)

The CRG20 has a discrete input which is capable of resetting the device. The RESET_IN input is an active-low signal. When pulled low for a minimum of 2 μ s, the ASIC in the CRG20 enters a reset state. If the RESET_OUT and RESET_PROC pins are connected, the ASIC will then also reset the processor in the CRG20. The timing of the reset sequence is shown in Figure 12.

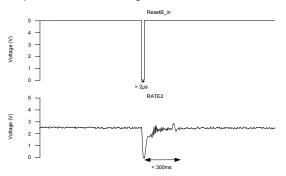


Figure 12 - RESET_IN Timing

RESET_IN Logic Levels

RESET_IN = 0 After a minimum of 2 μ s, the CRG20 is held in reset.

RESET_IN = 1 The CRG20 is in normal operation.

Auxiliary Input Channels

CRG20 has a pair of Auxiliary Analogue Inputs that are available to the user. These inputs are converted into digital format synchronised to the operation of the gyro. The data can be accessed as single packets for each input together with rate information.

The measurements can be read by the host system using the "Rate / Aux 1" and "Rate / Aux 2" message types (see page 11).

There are three constraints that the user needs to be aware of:

1. To maximise the fault detection capability of the sensor system, the inputs are checked for underflow and overflow. Therefore, it is essential to constrain the input signal between $V_{SSA} + 0.25$ V to $V_{DD5V} - 0.25$ V. Failure to do so will be interpreted as an invalid input signal and result in a logic '1' in bit 4 of the Status byte. (**BF**: Built in test (BIT) fault).

2. The input impedance of the auxiliary channels lies between 50 k Ω and 300 k Ω , with up to 10 pF of input capacitance. It is therefore necessary to ensure that the input is fed from a source that has sufficiently low output impedance that the signal is not adversely attenuated by the aux input itself.

3. The Aux inputs are sampled and converted for a maximum data rate of 1 kHz. Therefore, the input signal must be constrained to not exceeding 500 Hz to prevent aliasing.

Analogue Rate Output Channels

There are two analogue rate outputs from the CRG20, RATE1 and RATE2. Both rate outputs are referenced to VREF1, bipolar (around VREF1) and directly proportional to the applied rate.

RATE1 has a minimum voltage range of ± 2.1 V with respect to VREF1. When positive rate is applied, the voltage increases (within the rate limits of the gyro).

RATE2 is effectively the RATE1 output scaled by -0.9, giving it a minimum voltage range of ± 1.89 V with respect to VREF1. When positive rate is applied to the CRG20, the RATE2 output voltage decreases (within the rate limits of the gyro).

The smaller voltage range used for RATE2 allows the signal to be used to show BIT failure. When a BIT failure occurs (with the exception of CBIT being asserted), the RATE2 output falls to within 0.1 V of VSS. During normal operation, the decreased voltage range means the output will never reach this level.

Voltage References

The CRG20 has two voltage reference outputs.

VREF1 is the 0 °/s rate reference for the RATE1 and RATE2 analogue output channels. VREF1 is nominally (VDD5V – VSS) / 2. VREF1 has an offset error of no greater than \pm 14 mV.

VREF2 is used internally by the ASIC in the CRG20. It is nominally (VDD5V - VSS) / 3 and is only brought out to a pad as it requires decoupling (see RECOMMENDED LAYOUT).

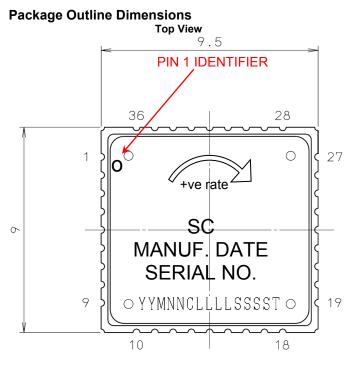
CRG20 Digital Angular Rate Sensor

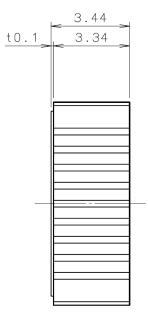


Page 12



MECHANICAL DESIGN





Side View

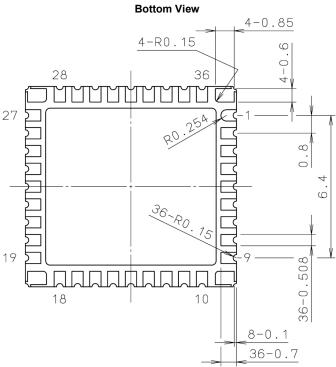


Figure 13 - Package Outline Dimensions



Lid Marking

Did Mutking			
Description	Code	Range	Necessary figures
② Right marking			
Logo	SSS		3
	(space)		1
Production type	CRG20		5
	-		1
Model number	NN	01	2
	•		1
Hardware configuration number	С	1	1
		Total figures	13
③Lower marking (=matrix marking code)			
Year number	ΥY	00-99	2
Month number	М	1-9,X,Y,Z	1
Model number	NN	01	2
(01:75deg/s,02:300deg/s)			
Hardware configuration number	С	1	1
Package (=MEMS) lot number	LLLL	0000-4095	4
Individual device serial number	SSSS	0000-9999	4
Measuring times	Т	0-9	1
		Total figures	15
(4) Top marking			
Production Country	Made in Japan		13
		Total figures	13
(5) Left marking			
Miyoshi marking	SD-01911003	9	
nigosii marking			

Figure 14 – Marking Detail

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Alignment & Tilt Specification

In order to ensure the cross-axis sensitivity specification is met, the CRG20 should be mounted parallel to the board so that after soldering it has a tilt of 0.25° or less, with respect to the mounting surface.

Recommended Solder Conditions

The CRG20 is designed to be reflow soldered to a PCB. It is recommended that lead-free solder flow conditions should not exceed 235 °C with a dwell over 220 °C of less than 60 seconds. However, short duration dwells of 10 seconds or less at temperatures of up to 260 °C can be tolerated.

Package Details

The CRG20 package is made from alumina high temperature co-fired ceramic (HTCC).

The pads on the CRG20 are electroplated with 5 μm of nickel, followed by up to 0.5 μm of electroplated gold.

Product Environmental Management Compliance

The CRG20 complies with the requirements of the RoHS directive.





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CRG20 Digital Angular Rate Sensor

Page 16

CRG20-00-0100-110 Rev 13 DCR710016657

